



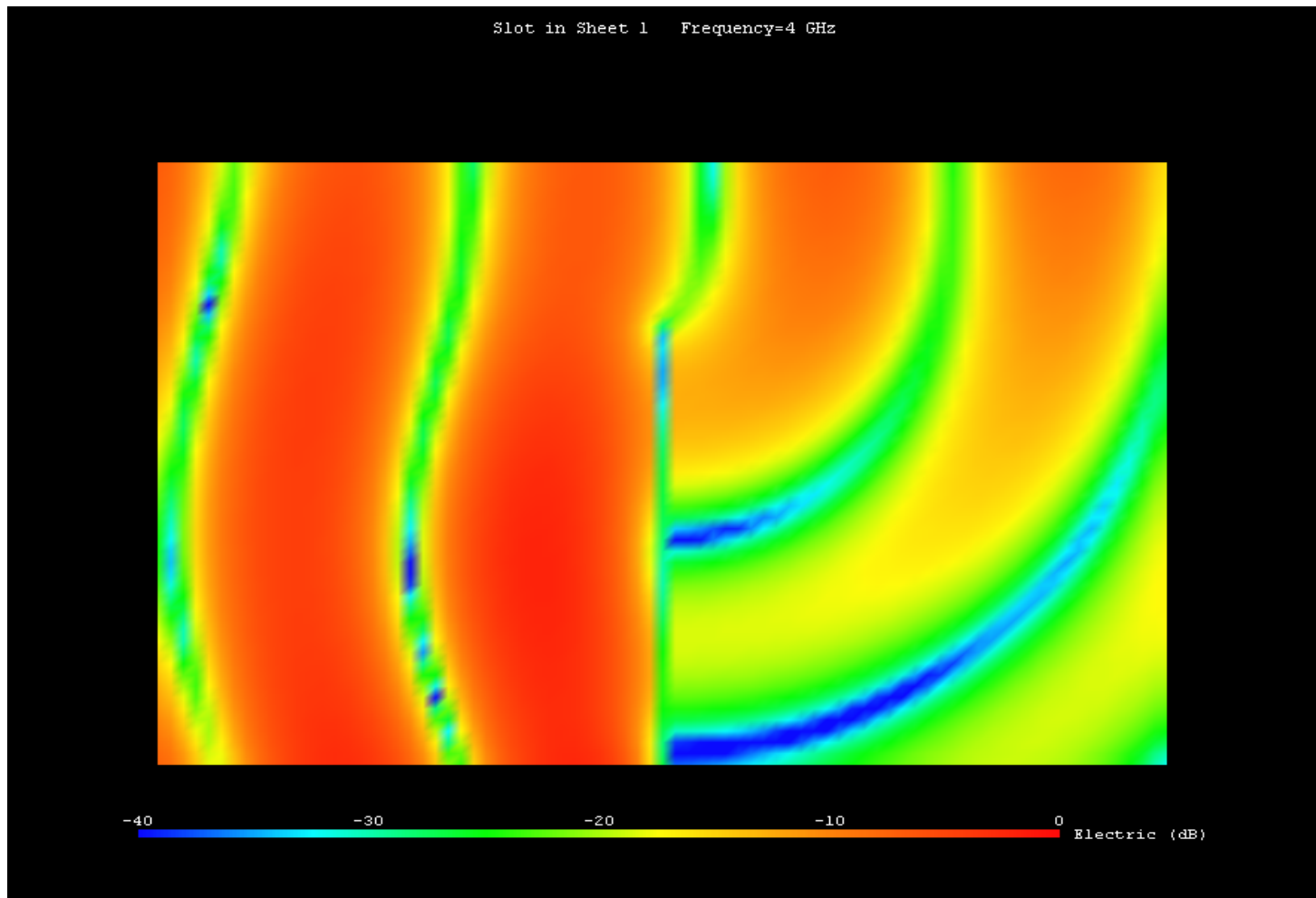
# Screening

Keeping the “baddies” out

# Scope of Topic

- A screen reduces transmission of Electric or/and Magnetic Fields
- Complete blocking is unlikely
- Effectiveness depends on wave impedances
  - Near field/far field
- EM fields are not blocked completely by a partial shield

# Partial Shield



# What Happens at an interface?

- Some energy is transmitted
- Some energy is reflected
- If edges exist, energy will creep round them
- If slots, or holes, exist energy will flow through them
  - Dependent on dimensions and frequency

# Topics

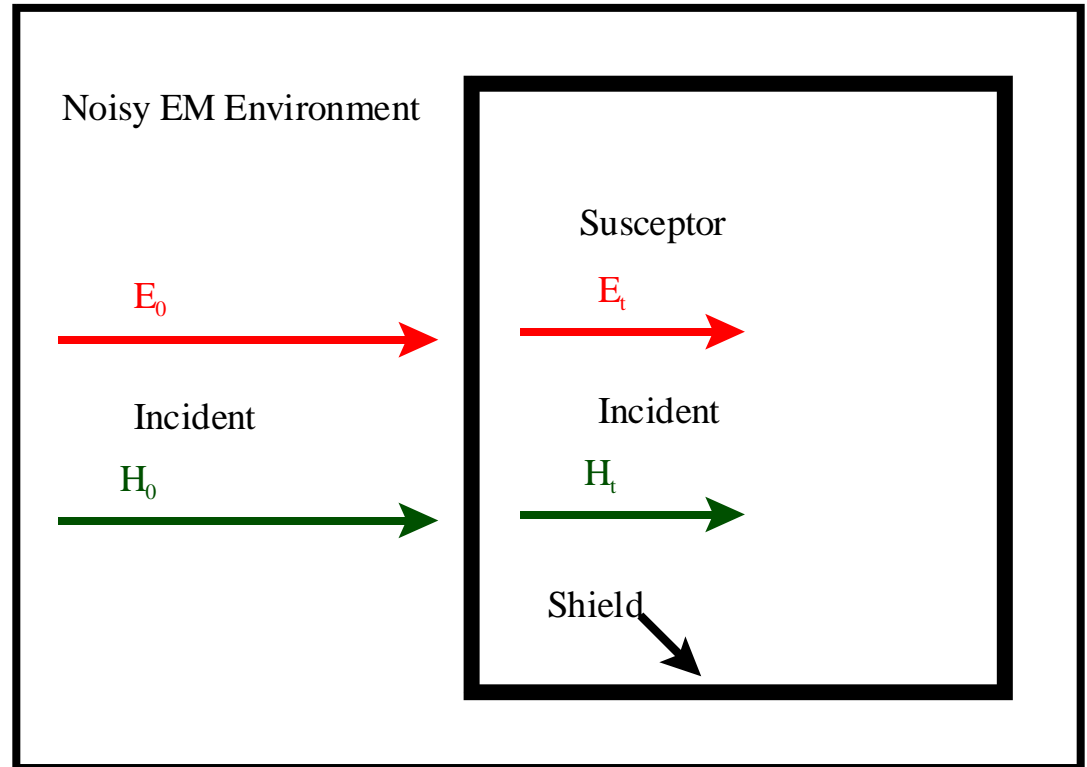
- Consider hermetic sealed containers
- Look at apertures and their behaviour
- Produce simple equations to model their behaviour
  - Shielding Effectiveness

# Hermetic Shield

## ■ Shielding Effectiveness

$$S = 20 \log \left( \frac{E_0}{E_t} \right) \quad dB$$

$$S = 20 \log \left( \frac{H_0}{H_t} \right) \quad dB$$



# Shield Materials and Mechanisms

- Partial reflection at the surface
  - Reflection loss
- Internal losses
  - Attenuation
- Generally require magnetic material to shield magnetic fields
- Eddy currents can be used in non-magnetic materials
  - Require high frequencies

# Shield Materials and Mechanisms

- Purely conductive shields shield
  - Electric Fields
    - Partial reflection at surface
    - Absorption
    - Multiple reflections
  - Magnetic fields at high frequencies –
    - IF the screen is thick



# Total Shielding Effectiveness

$$S = R + A + B \quad dB$$

- R is Reflection Loss
- A is Absorption Loss
- B accounts for internal reflections
  - Ignore if  $A > 10dB$

# Shielding Effectiveness

- Depends on Wave Impedance and Characteristic Impedance of Shield
  - Require an impedance mismatch
- Generally, for any medium

$$Z = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\varepsilon}}$$

- where the conductivity

$$\sigma = \sigma_r \sigma_{Cu}$$

# Shielding Effectiveness

■ Insulators  $\sigma \ll \omega\epsilon$  so  $Z = \sqrt{\frac{\mu}{\epsilon}} = 377\Omega$

■ Good Conductor  $\sigma \ll \omega\epsilon$

■ Shield Impedance

$$|Z| = \sqrt{\frac{\omega\mu}{\sigma}}$$

$$\mu_0 = 4\pi \times 10^{-7} \text{ Hm}^{-1} \quad \sigma_{Cu} = 5.28 \times 10^7 \text{ Sm}^{-1}$$

# Some Common Screening Materials

Material	$\sigma_r$	$\mu_r$	$ Z_s $ ( $\Omega$ )
Copper	1	1	$3.68 \times 10^{-7} \sqrt{f}$
Aluminium	0.61	1	$4.71 \times 10^{-7} \sqrt{f}$
Steel	0.1	1000+	$3.68 \times 10^{-5} \sqrt{f}$

# Absorption Loss

- TEM wave in a lossy medium loses amplitude exponentially

$$E = E_1 e^{-\frac{x}{\delta}} \quad H = H_1 e^{-\frac{x}{\delta}}$$

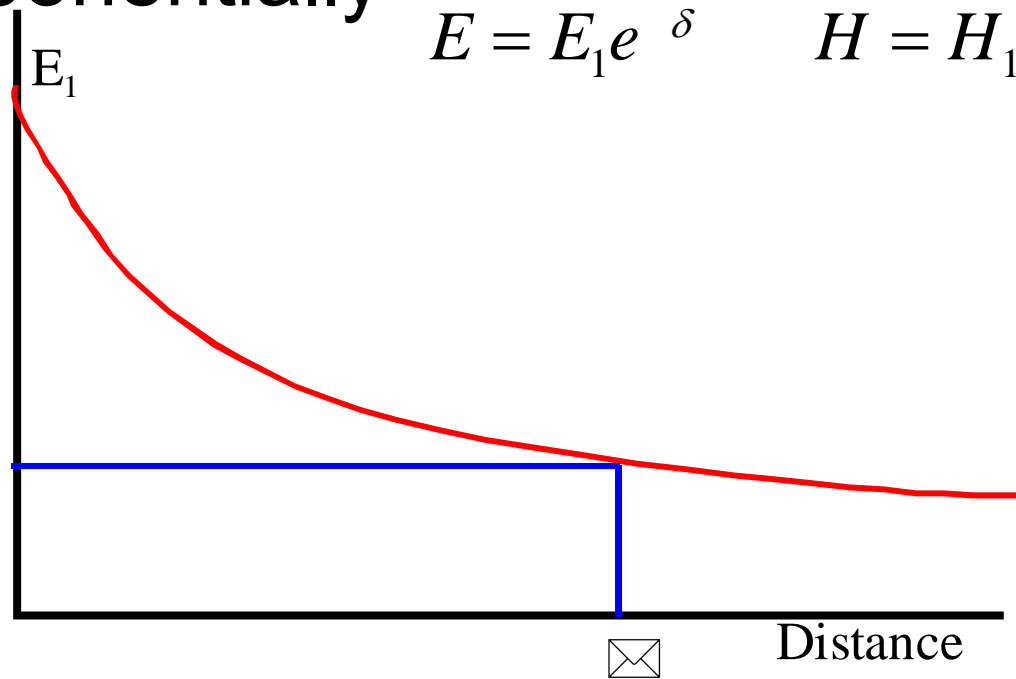
- Skin Depth

$$\delta = \sqrt{\frac{2}{\omega\mu\sigma}}$$

$m$

$E_1/e$

- $1/e = 36.8\%$



# Typical Values for Skin Depth (cm)

Frequency			
50Hz	0.933	1.195	0.12
1kHz	0.201	0.269	0.027
1MHz	0.0066	0.0085	0.0009

# Absorption Loss through a Shield

- Express  $E_t$  in terms of the exponential reduction

$$A = 20 \log \frac{E}{E_1}$$

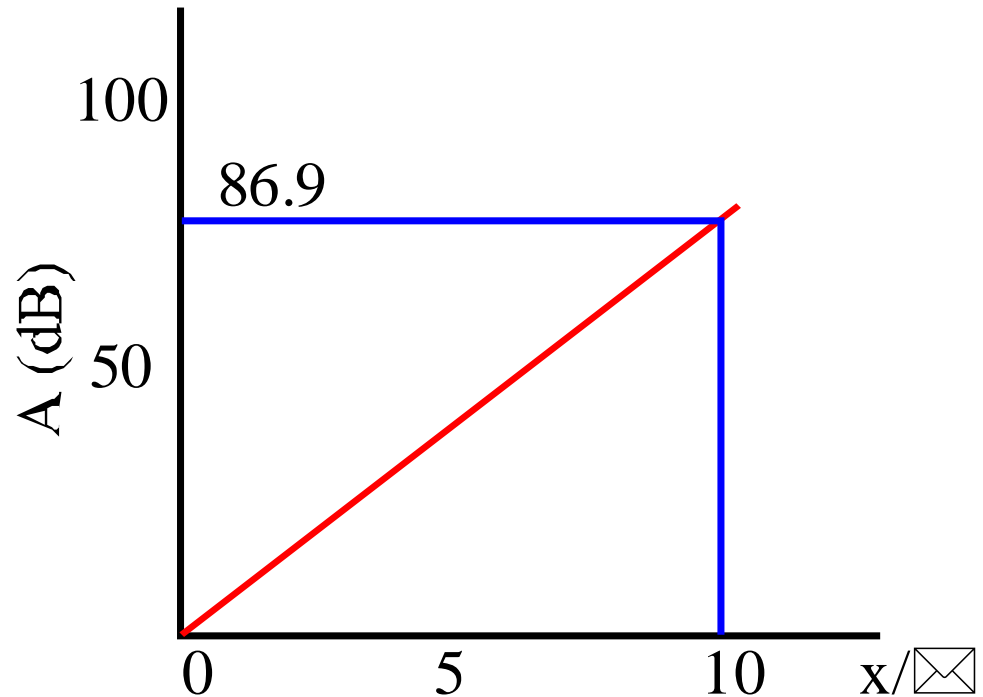
$$A = 20 \log \left( e^{-\frac{x}{\delta}} \right)$$

$$A = 20 \frac{x}{\delta} \log e$$

# Absorption Loss through a Shield

$$A = -8.69 \frac{x}{\delta} \quad dB$$

- Loss is proportional to plate thickness
- True for E and H





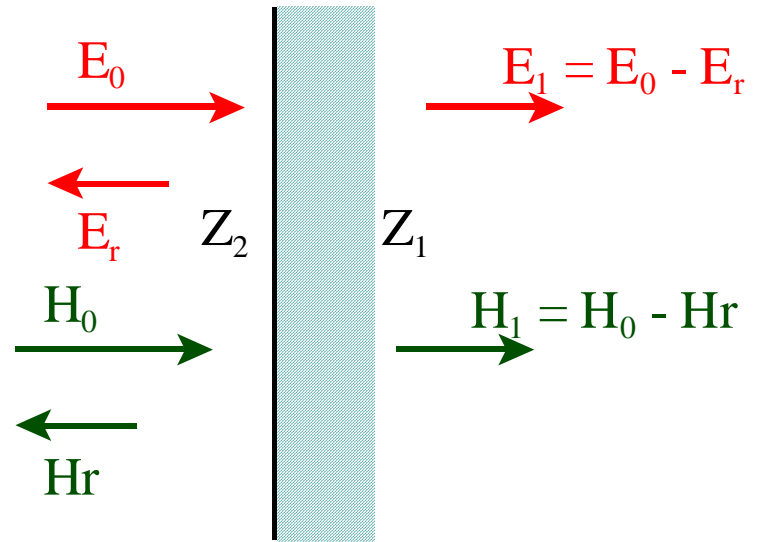
# Reflection Loss

- Relates to Reflection Coefficient.
- Normal Incidence

$$\frac{E_r}{E_0} = \frac{Z_1 - Z_2}{Z_1 + Z_2}$$

$$\frac{E_1}{E_0} = \frac{2Z_2}{Z_1 + Z_2}$$

$$\frac{H_1}{H_0} = \frac{2Z_1}{Z_1 + Z_2}$$



# Reflection Loss Through a Plate

## ■ Two Boundaries

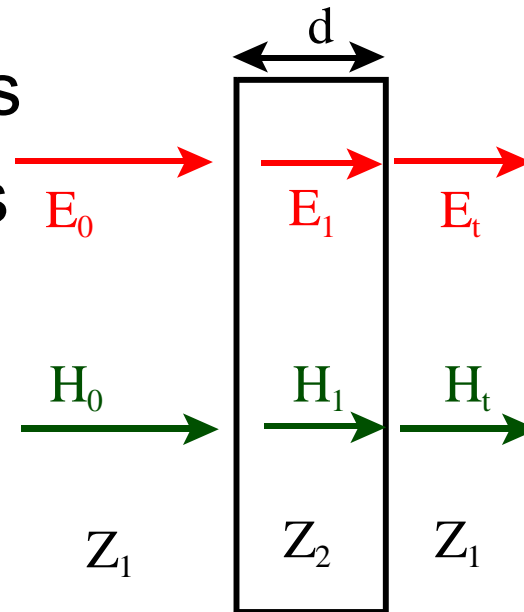
□ Combine two equations

□ Ignore Absorption Loss

$$\frac{E_t}{E_1} = \frac{2Z_1}{Z_1 + Z_2}$$

$$\frac{E_t}{E_0} = \frac{4Z_1Z_2}{(Z_1 + Z_2)^2}$$

$$\frac{H_t}{H_0} = \frac{4Z_1Z_2}{(Z_1 + Z_2)^2}$$



# Absorption Loss

- Model as an exponential

$$L_A = e^{-\gamma d}$$

- Only applies to second reflection loss term
- Exponential is used to account for multiple reflections in sheet

# Electric Fields

- Wave impedance is high
- Shield impedance is low
- Substituting  $Z_w$  and  $Z_s$  gives

$$\frac{E_t}{E_0} = \frac{4Z_w Z_s}{(Z_w + Z_s)^2} \approx \frac{4Z_s}{Z_w}$$

$$R = 20 \log \left( \left| \frac{Z_w}{Z_s} \right| \right) \quad dB$$

# Plane Waves

- Similar picture to E fields
- Wave Impedance is  $377\Omega$
- Shield Impedance is very low
- In both cases
  - Virtually all reflection occurs at first interface
  - Thin conducting shield is normally sufficient

# Magnetic Fields

- Wave Impedance is low
- Shield Impedance is low
- Reflection from interface is low
- Require absorption to attenuate magnetic fields
- Usually use magnetic materials with high permeability to increase the shield impedance

# Composite Fields

- Use twin shield to solve the problem

- Magnetic Shield

- High permeability

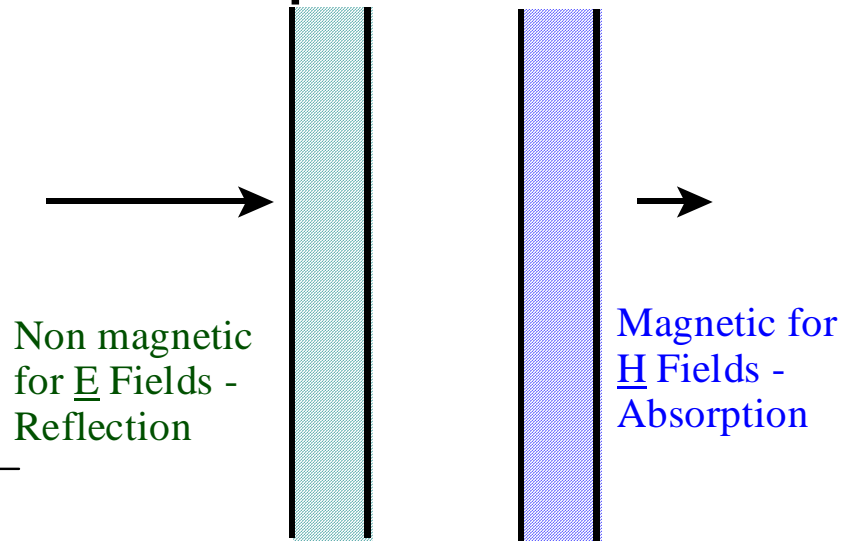
- Increases  $Z_s$

$$Z_s = \sqrt{\frac{\omega\mu}{\sigma}}$$

$$\delta = \sqrt{\frac{2}{\omega\mu\epsilon}}$$

- Reduces  $\delta$

- Increases Absorption



# Seams and Holes

- Always present
  - Access
  - Cables
  - Ventilation
- Leakage dominates propagation through the case
- Concentrate on the seams and holes

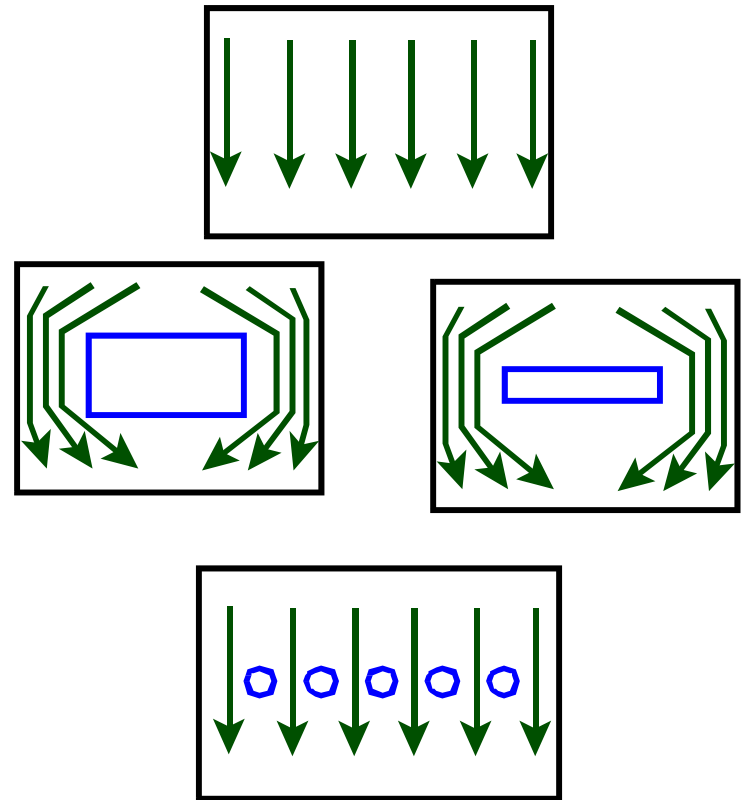


# Seams and Holes

- Magnetic fields are better “escape artists”
  - They leak more through seams and holes
- Leakage depends on
  - Maximum linear dimension of the hole
  - Aperture wave impedance
  - Frequency of the source

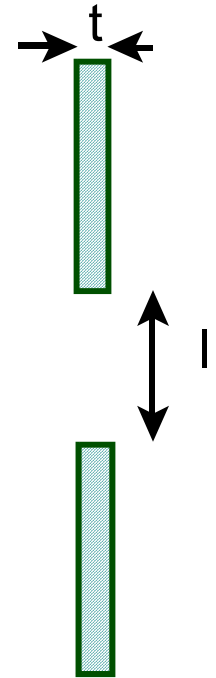
# Shielding Mechanism

- Eddy currents in shield
- Undistorted
- Slots distort currents
- Holes improve the problem



# Slot Antenna

- Allows development of simple method for handling slots/holes
- Applies to thin shields  $t < l$
- Efficient radiator for  $t = \lambda/2$
- For narrow slots the perimeter is  $2\lambda$ .



# Circular Aperture

- Treat as an aperture antenna

$$G = \frac{4\pi A}{\lambda^2} = \left( \frac{2\pi r}{\lambda} \right)^2$$

- Substituting for the area
- Magnetic fields tend to penetrate the aperture

# Shield Effectiveness

- $P_1$  is power transmitted with no shield present
- $P_2$  is power transmitted with shield present ( including hole)

$$S = 10 \log \left( \frac{P_1}{P_2} \right) \quad dB$$

- Gain is  $P_2 / P_1$  so  $S = 20 \log \left( \frac{\lambda}{2\pi r} \right) \quad dB$ 
  - Circular hole
  - Reflection loss at the hole

# Multiple Holes

- For  $n$  holes

$$G \approx n \left( \frac{2\pi r}{\lambda} \right)^2$$

- giving

$$S = 20 \log \frac{1}{\sqrt{n}} \left( \frac{\lambda}{2\pi r} \right) \quad dB$$

# Narrow Slot

- Leakage related to maximum length

$$S = 20 \log \left( \frac{\lambda}{2l} \right) \quad dB$$

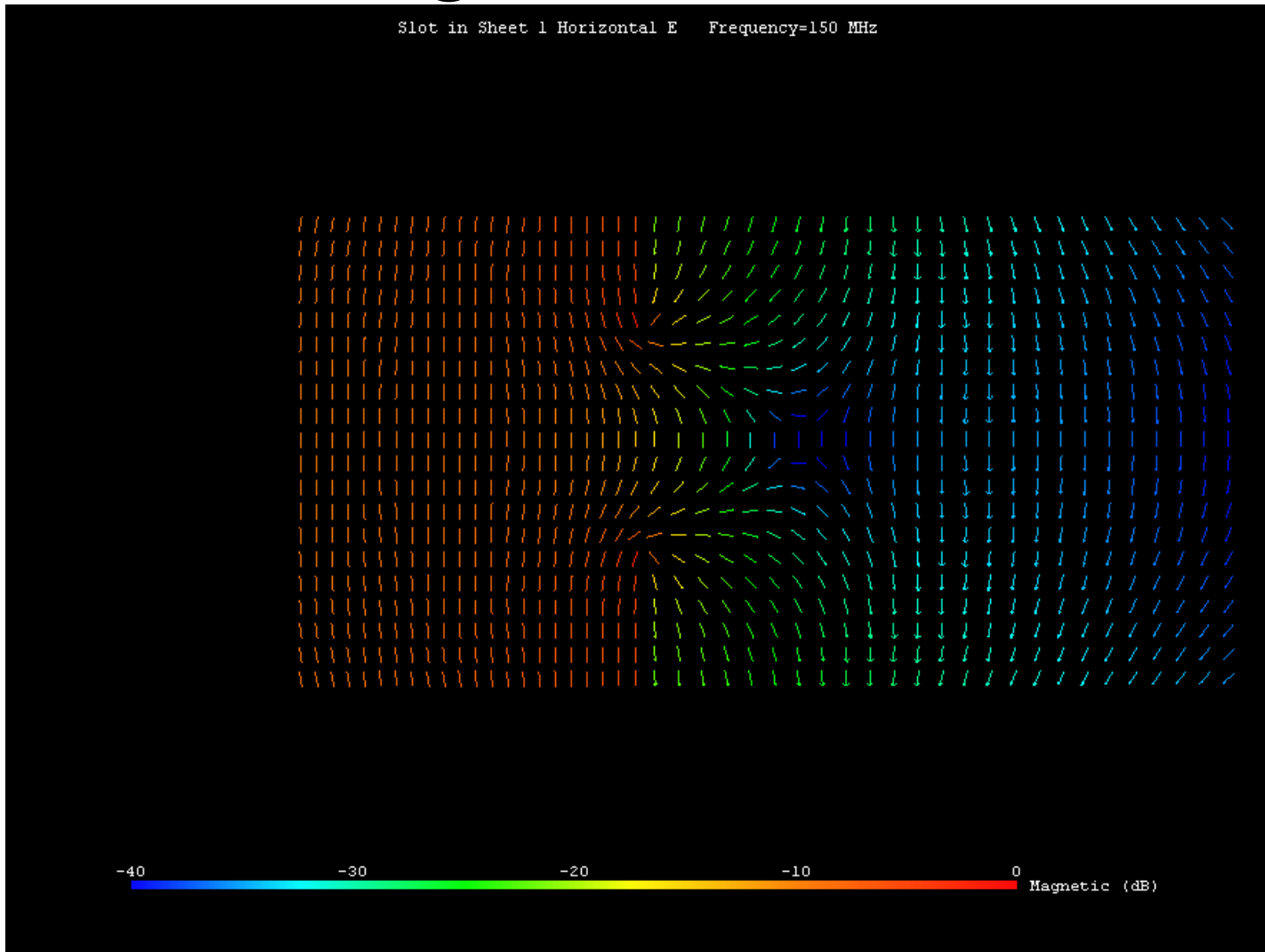
- For thin shields
  - $l < \lambda/50$
- $S = 28dB$
- Power ratio is 630

# Rectangular Slot

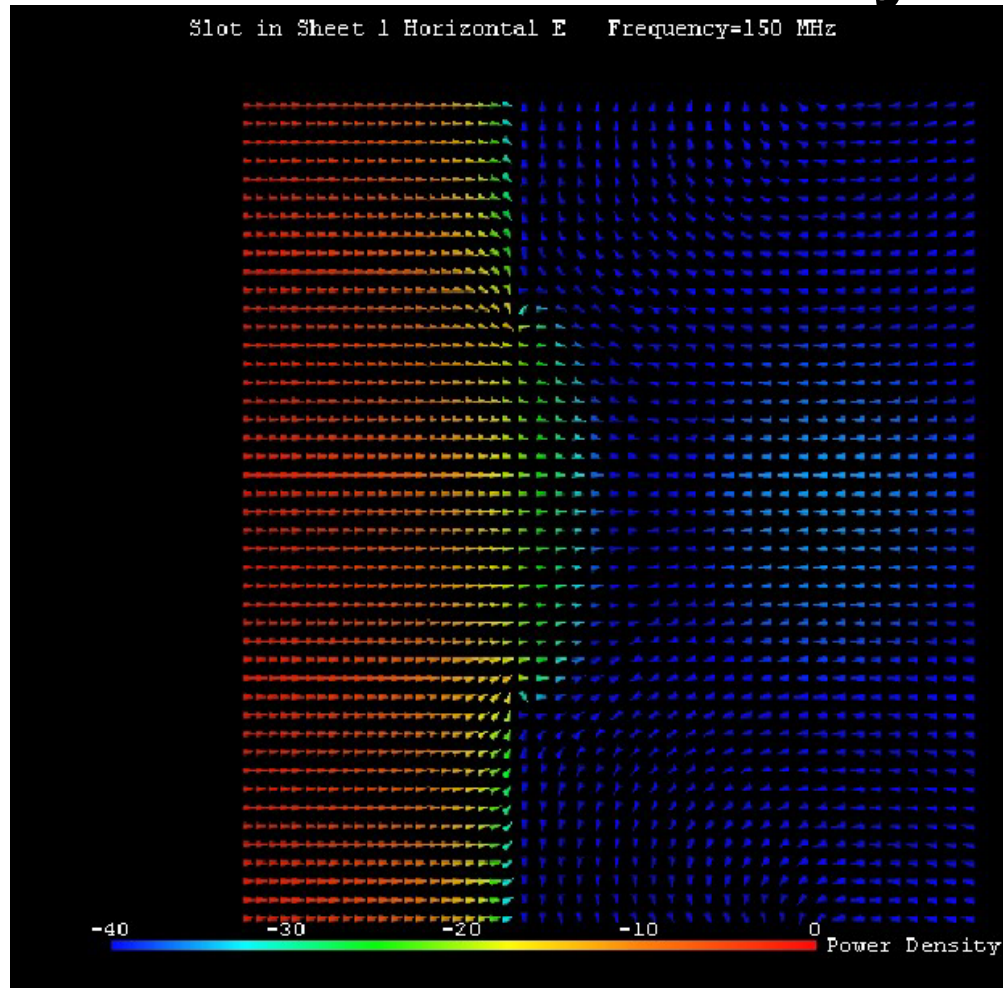
- 40mm x 20mm
- E field aligned with 40mm side
- 3.7GHz resonant frequency
- Fields at 150MHz, 2.4GHz and 14GHz



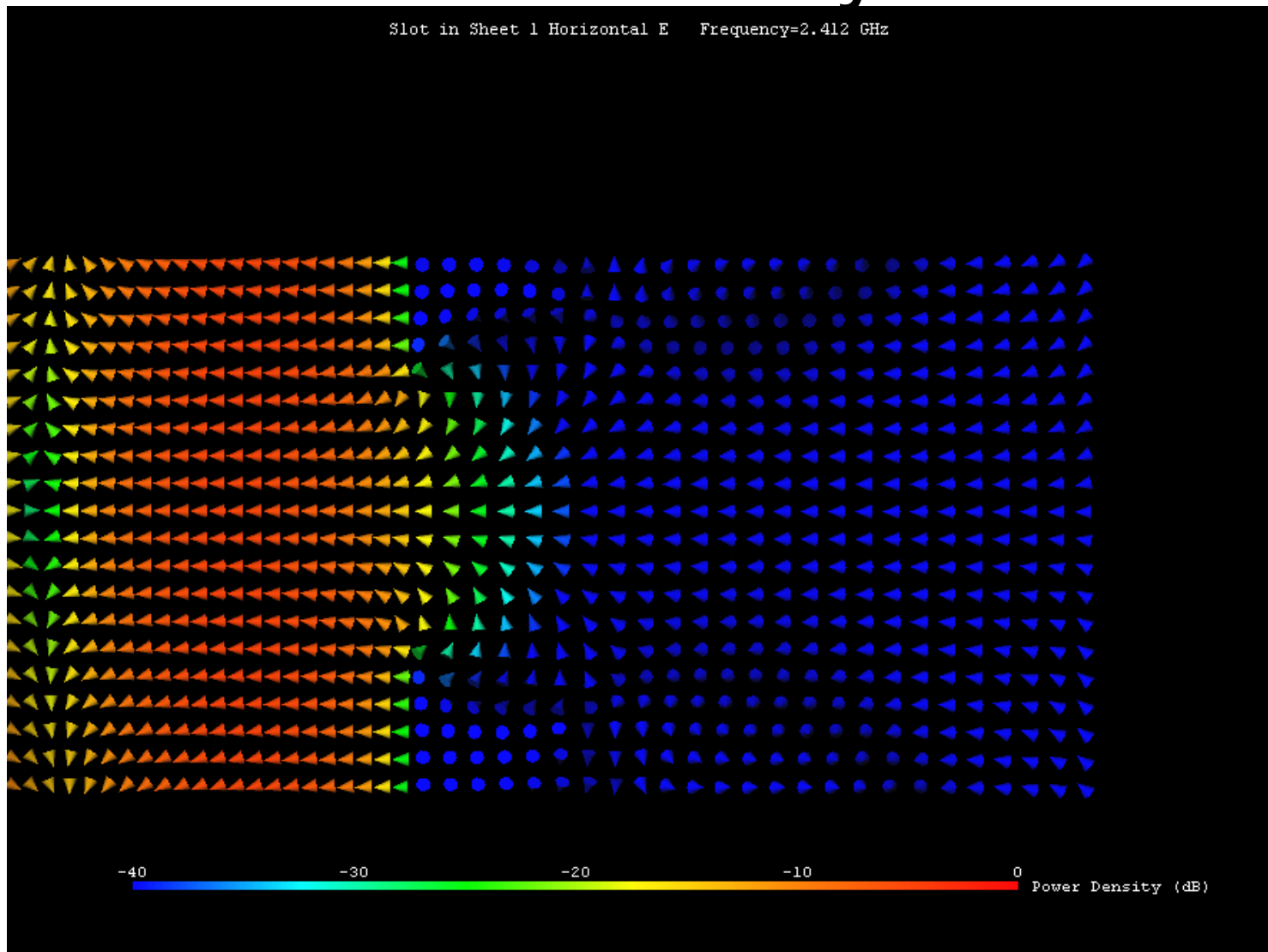
# 150MHz Magnetic Field



# 150MHz Power Density

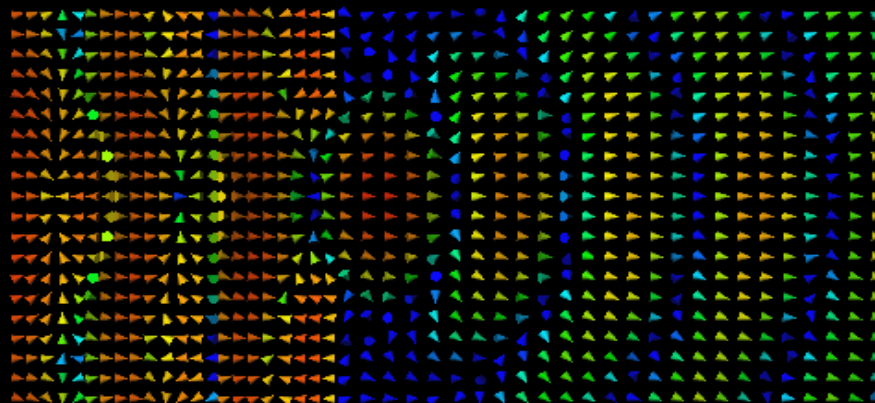


# 2.4GHz Power Density



# 14GHz Power Density

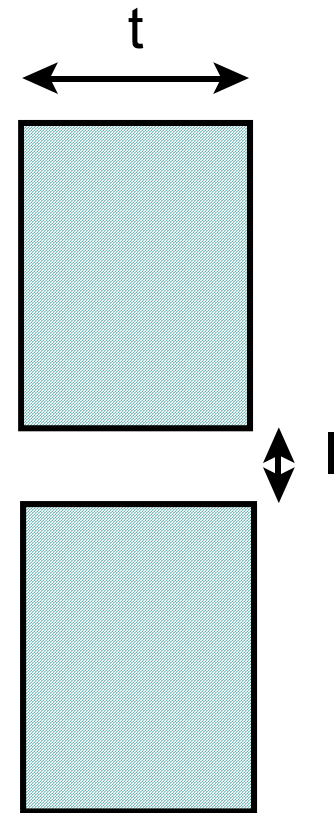
Slot in Sheet 1 Horizontal E Frequency=14 GHz



-40 -30 -20 -10 0 Power Density (dB)

# Thick Shields

- Defined by  $t > l$
- Aperture acts as a waveguide
- If aperture dimension is less than  $\lambda/2$  screening will be good
- Cut-off wavelength for narrow slot  $\lambda_c = 2l$
- Cut-off is not complete



# Thick Shields

- Cut-off frequency in air

$$f_c = \frac{c}{\lambda_c} = \frac{1.57 \times 10^7}{l}$$

- Below  $f_c$  the mechanism is waveguide loss

- Attenuation of the form

$$H = H_0 e^{-\alpha x}$$

- For  $f \ll f_c$

$$\alpha \approx \frac{\omega}{c} \frac{f_c}{f}$$

# Thick Shields

- Shielding effectiveness

$$S_{wg} = 20 \log \left( \frac{H_{entry}}{H_{exit}} \right) = 20 \log(e^{\alpha t})$$

$$S_{wg} = 20 \alpha t \log(e) = 8.69 \alpha t \quad dB$$

- Note similarity to absorption loss in solid plate

$$S_{wg} \approx 8.69 \frac{2\pi f}{c} \frac{c}{2l} \frac{1}{f} t = 27.2 \frac{t}{l} \quad dB$$

# Thick Shields

## ■ Note

□ frequency independence

□  $f \ll f_c$

## ■ Total Shielding Effectiveness

$$S = 20 \log \left( \frac{\lambda}{2l} \right) + 27.2 \frac{t}{l} \quad dB$$

Reflection

Absorption



# Thick Shields

$$\lambda_c = 1.7d$$

- Circular Hole  $t > d$

$$f_c = \frac{c}{\lambda_c} = \frac{1.75 \times 10^7}{d} \quad \text{in air}$$

$$\alpha = \frac{\omega}{c} \frac{f_c}{f} = \frac{2\pi f}{c} \frac{c}{1.7d} \frac{1}{f} = \frac{3.7}{d}$$

- giving

$$S_{wg} = 8.69 \frac{3.7}{d} t = \frac{32t}{d} \quad \text{dB}$$

# Thick Shields

- Total Shielding Effectiveness

$$S = 20 \log \left( \frac{\lambda}{\pi d} \right) + \frac{32t}{d} \quad dB$$

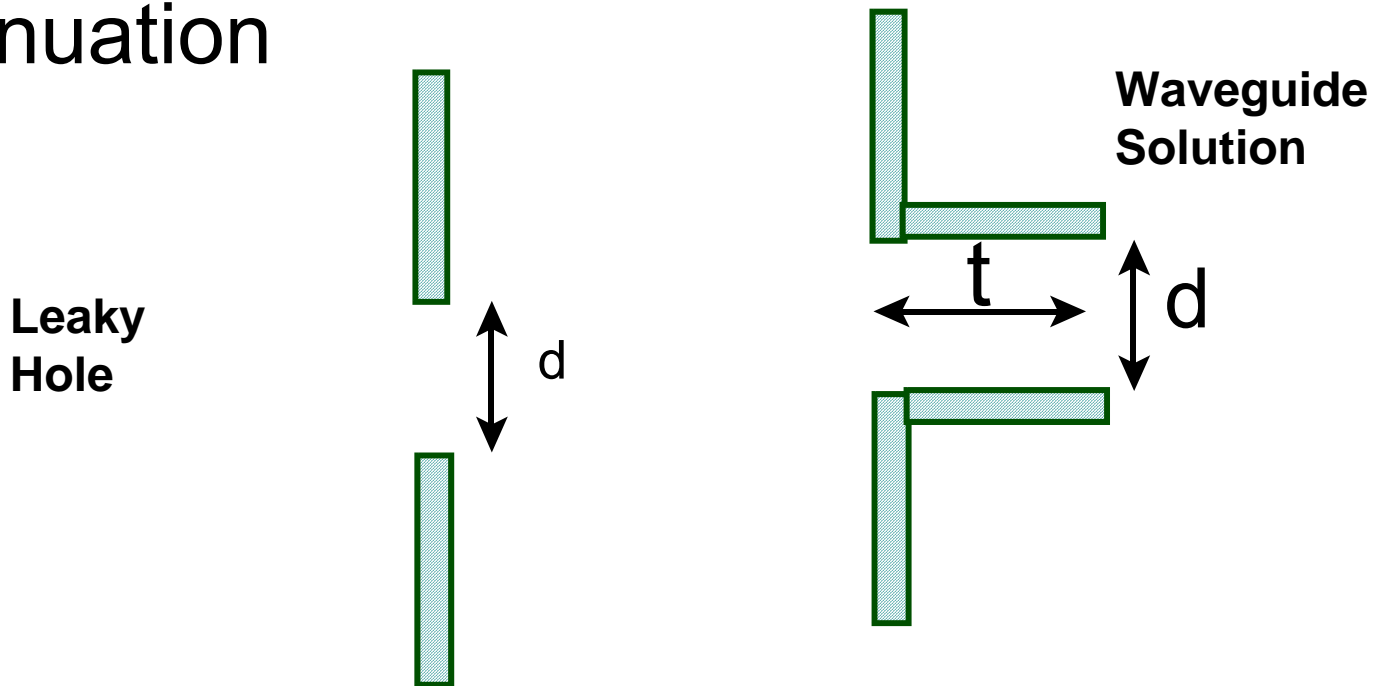
Reflection

Absorption

- If  $\lambda = 5\lambda_c$  and  $t = 2d$
- $S = 8.65 + 64 \text{ dB}$
- Waveguide absorption is far more effective

# Handling a hole for cable entry

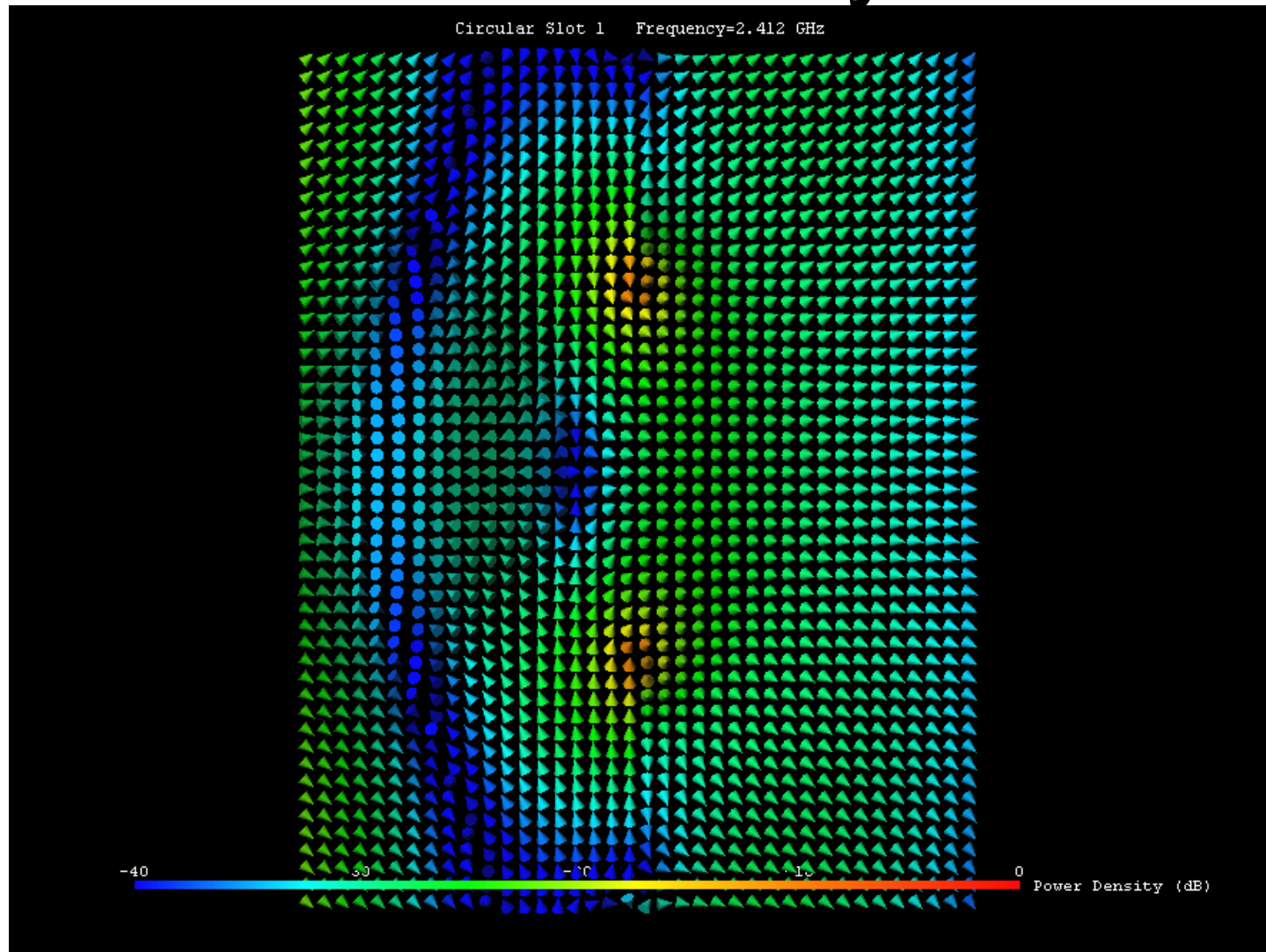
- The best technique is to increase shielding effectiveness by adding waveguide attenuation



# Circular Apertures

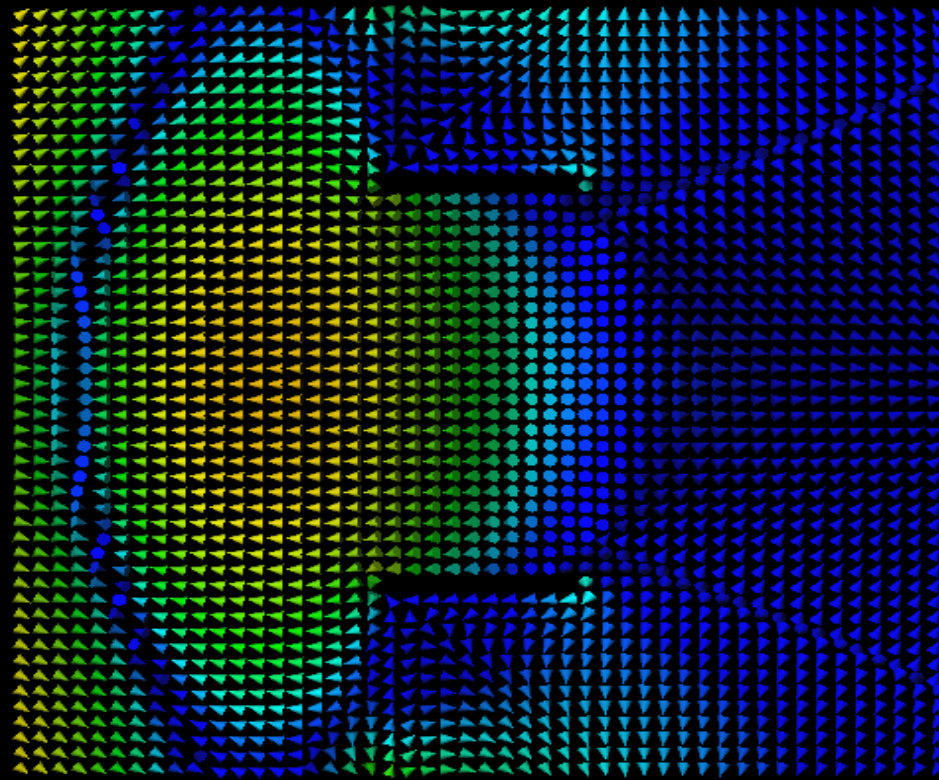
- Radius 25mm
- Radius 25mm with 20mm tube extension

# 2.4GHz Power Density



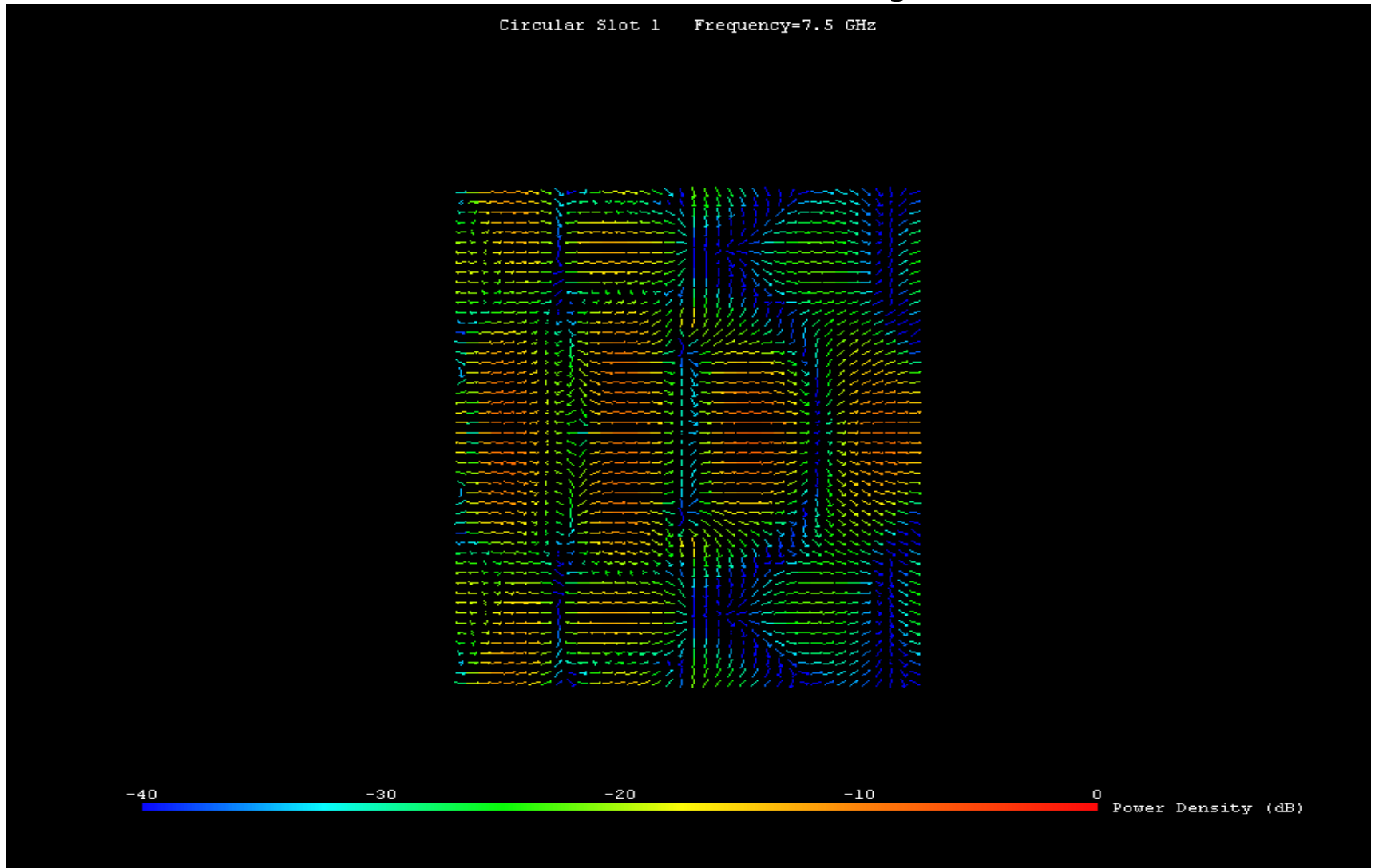
# 2.4GHz + 20mm tube

Circular Slot 1 Frequency=2.412 GHz

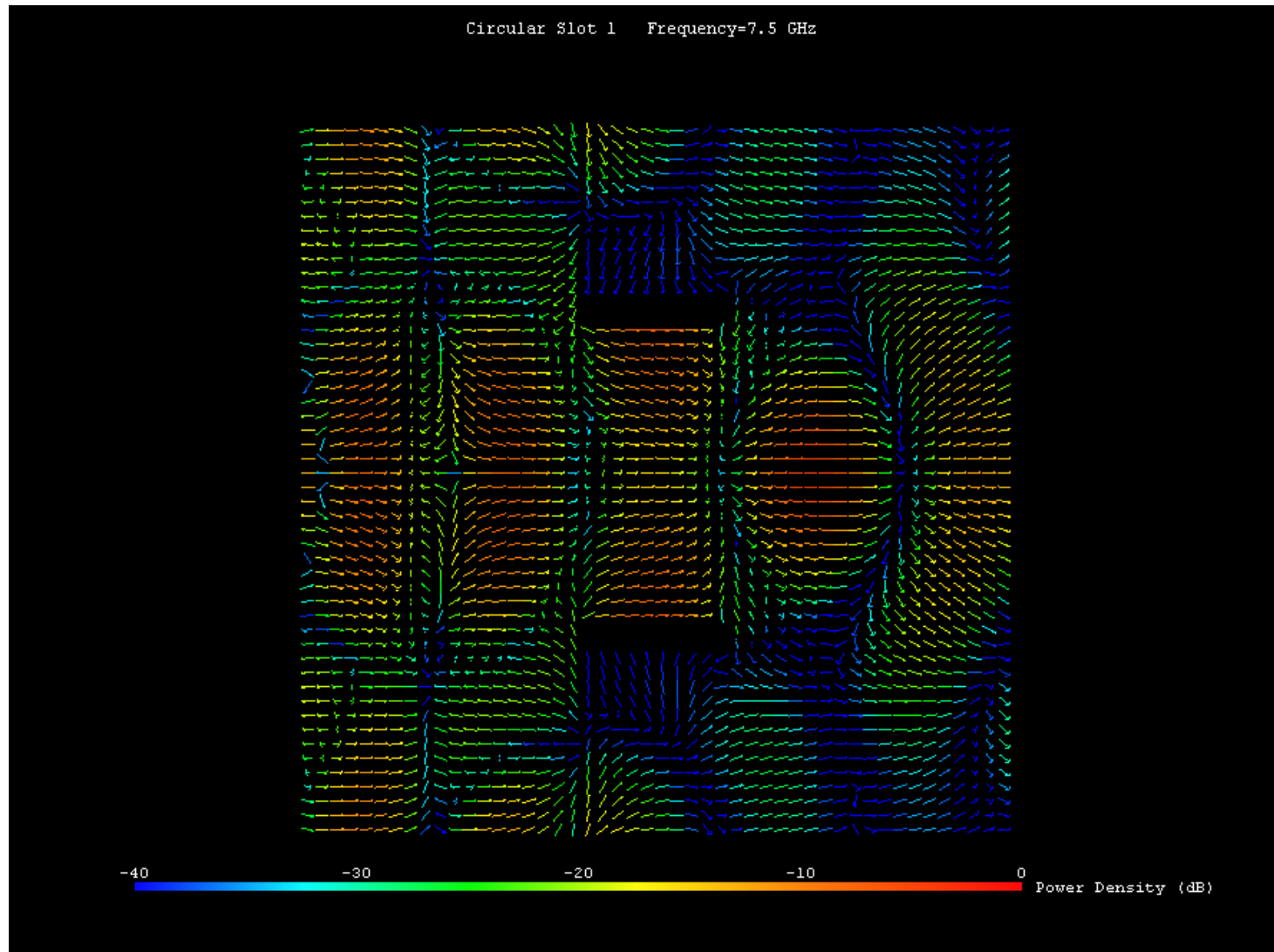


-40 -30 -20 -10 0 Power Density (dB)

# 7.5GHz Power Density



# 7.5GHz + 20mm tube







# Approach to the problem

- Find field strength at the shield
- Know what type of field and shield you have
- Apply shielding factor
- Apply inverse square law to get field strength some distance away from the shield

# Summary

- Shielding can be achieved for all types of EM fields
- Simple equations have been derived from the basic theory
- Holes and seams cause significant problems with shielding